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Operating and Loading Conditions of Three-level Neutral-Point-Clamped Wind Power Converter under Various Grid Faults

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Abstract – In order to fulfill the growing demands from the grid side, the full-scale power converters are becoming popular in the wind turbine system. The Low Voltage Ride Through (LVRT) requirements may not only cause the control problems but also result in overstressed components for the power converter. However the thermal loading of the wind power converter under various grid faults is still not yet clarified especially at MW power level. In this paper, the impacts by three types of grid faults to a three-level Neutral-Point-Clamped (3L-NPC) wind power converter in terms of operating and loading conditions are analytically solved and simulated. It has been found that the operating and loading of converter under LVRT strongly depend on the types/severities of grid voltage dips and also the chosen control algorithms. The thermal distribution among the three-phases of converter may be quite uneven and some devices are much more stressed than the normal operating condition.

Index Terms – Wind power generation, multilevel converter, LVRT, thermal loading.

I. INTRODUCTION

As a promising and fast growing renewable energy source, the accumulated capacity of wind power generation has achieved 238 GW globally in 2011 [1]. Meanwhile, the power capacity of individual wind turbine is also increasing continuously to reduce the price per produced kWh. In 2012, 8 MW wind turbines with diameter of 164 m have already been presented to the market [2]-[4], and the newly established wind turbines are mainly located at remote area. As a result, due to more significant impacts to the grid stability and high cost for maintenance/repair, the reliability and ability to withstand grid disturbances are greatly emphasized for the modern wind turbines.

The Transmission System Operators (TSO) in different countries have issued stricter Low Voltage Ride Through (LVRT) codes for the Wind Turbine Systems (WTS), as shown in Fig. 1 [5], in which the boundaries for dipping amplitudes of grid voltage and the disturbing time are defined. Moreover, it is becoming a need that the WTS should provide some reactive power (up to 100 % current capacity) to help the grid recover during voltage dips. Fig. 2 shows an example from German grid codes in which the required reactive current against to the amplitude of grid voltage is specified [6].

The stricter LVRT codes as well as the higher requirements for reliability push the solutions of WTS moving from Doubly Fed Induction Generator (DFIG) with partial-scale power converter to Asynchronous/Synchronous Generator with full-scale power converter [7]. Intensive research have been devoted to the control and PLL strategies of wind power converter under the grid faults [8]-[12], and there is also some investigations regarding the thermal loading and modulation of converter under balanced LVRT condition [13], [14]. However the operating and loading states of converter under various LVRTs (including unbalanced faults) have not been comprehensively investigated. These uncertainties could result in uneven utilization of power devices, under-estimated rating of components, and compromised reliability performances for the wind power converters [13].

Consequently, this paper will focus on the loading states of a 10 MW full-scale 3L-NPC converter under various
LVRT conditions. First, the impacts of various grid faults to the AC bus voltages and the converter operations are simulated and analytically solved. Afterwards the corresponding loss and thermal behaviors of the power semiconductor devices in the given wind power converter will be presented.

II. VOLTAGE DIPS UNDER VARIOUS GRID FAULTS

When a short-circuit fault happens in the power grid system, normally the voltage dips will be detected by the grid-connected converter through the AC bus it is connected, and then the corresponding LVRT control algorithm of the Wind Turbine System (WTS) is activated, making the converter shift from the normal operation mode to the LVRT mode. However, depending on the types and locations of the short circuit, the line impedance and configuration of transformer windings, the voltage dips may vary significantly on different AC buses in the power grid [15], [16]. Therefore, it is important first to investigate how the voltage dip looks like on the bus where the WTS is connected.

A typical configuration of a WTS with the grid system is shown in Fig. 3, in which the voltage on Bus 2 is monitored by the WTS and hence determines the LVRT behavior of the wind power converter. A delta-star transformer is used to interface the WTS on Bus 2 (e.g. 3.3 kV) and the Point of Common Coupling (PCC) on Bus 1 (e.g. 20 kV). Other power sources and loads in the distribution system may be also connected to the Bus 1. It is assumed that a short-circuit fault happens somewhere with line impedance \( Z_F \) to the Bus 1 (PCC), and the line impedance from PCC to the grid with higher voltage level is \( Z_S \).

Define that the voltage dip severity on Bus N is \( D_N \), which is related to the location of grid faults and power line impedance. Providing that the line impedances for the positive and negative sequence components are equal, the dip severity on the Point of Common Coupling (or Bus 1) \( D_1 \) can be written as [16]:

\[
D_1 = \frac{Z_F}{Z_F + Z_S}
\]  

(1)

It is obvious that \( D_1 \) ranges from 0 (if \( Z_F = 0 \)) to 1 (if \( Z_S = 0 \)), and represents the voltage dipping severities from the most severe case \( (D_1 = 0) \) to the non-dip case \( (D_1 = 1) \).

Three typical grid faults: one-phase grounded (1 phase), two-phase connected (2 phase) and three-phase grounded (3 phase) are assumed to happen respectively at the same location of the power grid. Due to the delta-star connection of the transformer, the voltage dips on Bus 1 may have different characteristics when propagated on Bus 2. The dip type, dip severity and voltage amplitude of different grid faults are summarized in Table I, in which the voltage parameters seen on Bus 1 and Bus 2 are identified respectively. The voltage dipping type A-D is defined as phasor diagrams in Fig. 4 [15], [16].

Table I. Voltage dip seen on Bus 1 and Bus 2 for various grid faults.

<table>
<thead>
<tr>
<th>Fault type</th>
<th>One-phase</th>
<th>Two-phase*</th>
<th>Three-phase</th>
</tr>
</thead>
<tbody>
<tr>
<td>Dip type</td>
<td>B</td>
<td>C</td>
<td>A</td>
</tr>
<tr>
<td>Dip value</td>
<td>( D_1 )</td>
<td>( D_2 ) ( \angle 0^\circ )</td>
<td></td>
</tr>
<tr>
<td>Voltage amplitude ( V_{\text{min}} ) (p.u.)</td>
<td>( D_1 )</td>
<td>0.5\sqrt{3} ( D_2 ) ( \angle 0^\circ )</td>
<td>( D_1 )</td>
</tr>
<tr>
<td>Dip type</td>
<td>C</td>
<td>D</td>
<td>A</td>
</tr>
<tr>
<td>Equivalent dip value ( D_2 )</td>
<td>( D_2 + 0.5 \sqrt{3} \angle 0^\circ )</td>
<td>( D_2 ) ( \angle 0^\circ )</td>
<td>( D_2 ) ( \angle 0^\circ )</td>
</tr>
<tr>
<td>Voltage amplitude ( V_{\text{min}} ) (p.u.)</td>
<td>0.5\sqrt{3} ( D_1 ) ( \angle 0^\circ )</td>
<td>( D_2 )</td>
<td>( D_2 )</td>
</tr>
</tbody>
</table>

Notes:
* Two-phase grounded fault is not included because it seldom happens.
** No phase jump is assumed, i.e. vector \( D_1 \) has zero phase angle.
*** Voltage amplitude means the lowest one among the three phases.

![Fig. 4. Phasor diagram definitions for the dip types A-D specified in Table I.](image)

![Fig. 5. Minimum voltage amplitude on Bus 2 vs. Bus 1.](image)
The relationship between the minimum voltage amplitude among three-phases on Bus 1 \( V_{1\text{min}} \) and the minimum voltage amplitude on Bus 2 \( V_{2\text{min}} \) are plotted in Fig. 5, in which balanced (three-phase) and unbalanced (one-phase and two-phase) grid faults are indicated respectively. It is noted that because Bus 2 is directly monitored by the wind power converter, only \( V_{2\text{min}} \) will determine the amount of reactive current injected into the power grid by converter.

### III. LVRT OPERATION OF GRID-CONNECTED CONVERTER UNDER VARIOUS GRID FAULTS

In order to determine the loading of the power semiconductor devices, it is important to study how the grid-converter behaves under various grid voltage dips. Depending on the AC bus voltage and active/reactive power injection methods, the current and voltage amplitudes as well as their phase displacement of the grid converter may vary considerably and create different operation scenarios.

#### A. Converter system under investigation

As the widely commercialized multilevel topology which is used in the high-power medium-voltage applications for industry, mining, and traction [17], the three-level neutral-point-clamped (3L-NPC) solution seems to be a promising candidate for the Multi-MW full-scale wind power converter [18]-[20], as shown in Fig. 6. This topology is chosen and basically designed for a 10 MW wind turbine as a case study in this paper [19]-[22], where the major design parameters are summarized in Table II.

Fig. 7 shows the control scheme adopted to deal with unbalanced grid faults: both positive and negative sequences voltages/currents are detected and controlled. A sequence decoupling algorithm is used to remove the 100 Hz oscillation components in each of the sequence domain [11], [12]. It is worth to mention that the currents injection strategy under various grid faults is still an issue for discussion, in this paper it is assumed that the active and reactive current generated by the grid-converter only contains positive sequence component, and this can be achieved by setting the references for negative sequence currents to zero when using the control method shown in Fig. 7.

The references for positive sequence current as a function of the minimum voltage amplitude on Bus 2 \( V_{2\text{min}} \) are specified in Fig. 8, where the current \( I_q^+ \) is set according to the German grid code reported in Fig. 2, and the current \( I_d^+ \) is referred to the generated active power by wind turbines. It is assumed that the converter is set to provide as much active power as possible during the grid faults, and the pitch control does not have enough time to activate [8], [9]. This is the worst testing condition for the wind power converter.

### Table II. Parameters of three-level Neutral Point Clamped inverter for case study

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>Rated output active power ( P_o )</td>
<td>10 MW</td>
</tr>
<tr>
<td>DC bus voltage ( V_{dc} )</td>
<td>5.6 kV DC</td>
</tr>
<tr>
<td>Rated primary side voltage ( V_p )</td>
<td>3.3 kV rms</td>
</tr>
<tr>
<td>Rated line-to-line grid voltage ( V_g )</td>
<td>20 kV rms</td>
</tr>
<tr>
<td>Rated load current ( I_{load} )</td>
<td>1.75 kA rms</td>
</tr>
<tr>
<td>Carrier frequency ( f_c )</td>
<td>750 Hz</td>
</tr>
<tr>
<td>Filter inductance ( L_f )</td>
<td>1.1 mH (0.25 p.u.)</td>
</tr>
</tbody>
</table>

* Line-to-line voltage in the primary windings of transformer.
Table III. Voltage dips for example ($D_1=0.5$ p.u.).

* Voltage amplitude means the lowest one among the three phases.

Table IV. Positive and negative sequence voltage on Bus 2 for different fault types.

sequence voltage, there is a 100 Hz oscillation in the delivered active and reactive power under one-phase and two-phase unbalanced grid faults. This 100 Hz power oscillation is assumed to be absorbed by the DC bus chopper during LVRT operation.

C. Operation profiles with various dip severity and wind speeds.

After the bus voltage dips and corresponding converter behaviors are specified, it is possible to calculate the operating profiles of converter under various grid faults. In this paper the profiles for the delivered active/reactive power, phase displacement between current and voltage, and the modulation index are going to be investigated because they are closely related to the loading of switching devices.

If the grid voltage is aligned with the $d$ axis in the rotating coordinate, and the negative sequence currents are controlled to be zero, the active power $P$ and reactive power $Q$ delivered into the power grid are governed by:

$$
\begin{align*}
P &= \frac{3}{2} V_d^+ \cdot I_d^+ = \frac{3}{2} V_2^+ \cdot I_d^+ \\
Q &= \frac{3}{2} V_q^+ \cdot I_q^+ = \frac{3}{2} V_2^+ \cdot I_q^+
\end{align*}
$$

(1)

where the $d$ axis positive sequence voltage $V_d^+$ is equal to the positive sequence voltage on bus 2 $V_2^+$, which is given in Table IV for various grid fault conditions.

Then the positive sequence currents $I_d^+$ and $I_q^+$ generated by the converter can be calculated as:

$$
\begin{align*}
I_d^+ &= \sqrt{2} I_{\text{rated}} \cdot R_{\text{codes}} \\
I_q^+ &= \min \left[ \sqrt{2} I_{\text{rated}} \cdot \sqrt{1 - R_{\text{codes}}^2}, \frac{2 P_c(V_w)}{3 V_d^+} \right]
\end{align*}
$$

(2)

where $I_{\text{rated}}$ is the rated load current, $P_c(V_w)$ is the generated active power by the wind turbines, it is related to the wind speeds and the functions can be found in [23]. $R_{\text{codes}}$ is the reactive current ratio required by the German grid codes [6], it can be analytically represented as:
The modulation index for phase X (A, B or C) of converter $m_x$ can be calculated as:

$$m_x = \frac{\sqrt{2} \cdot V_{cs}}{V_{dc} / 2}$$  \hspace{1cm} (5)$$

where the converter output voltage $V_{cs}$ (phase to neutral point of three phases) can be calculated as:

$$V_{cs} = \sqrt{V_{gs} \sin(-\theta_s) + V_L^2 + V_{gs} \cos(-\theta_s)^2}$$  \hspace{1cm} (6)$$

where $V_{gs}$ represents the voltage amplitude of bus 2 in phase X, and the voltage drop on the filter inductance $V_L$ can be written as:

$$V_L = \omega L_f \sqrt{(I_d)^2 + (I_q)^2}$$  \hspace{1cm} (7)$$

It is noted that the modulation strategy will modify the duty ratio and thereby change the shape of voltage reference for converter (phase to mid-point of DC bus). The maximum modulation index will be extended from 1 to 1.15 if some modulation strategies with zero-sequence-components are applied.

Fig. 12 shows the modulation index of the grid connected converter at various dip severity on PCC ($D_1$). It can be seen that the modulation index is unsymmetrical among the three phases under the unbalanced fault conditions, and the wind speeds do not have strong impacts to the modulation index of converter under various LVRT operations.

Fig. 13 indicates the power oscillation amplitude under various dip severity for the three types of grid faults, and different wind speed conditions are indicated. The oscillation amplitude of active power $P_{cs2}$ and reactive power $Q_{cs2}$ can be calculated as [12]:

$$P_{cs2} = \sqrt{P_{\cos2}^2 + P_{\sin2}^2} = \frac{3}{2} \sqrt{(V_d^* \cdot I_d^*)^2 + (V_q^* \cdot I_q^*)^2}$$  \hspace{1cm} (8)$$

$$Q_{cs2} = \sqrt{Q_{\cos2}^2 + Q_{\sin2}^2} = \frac{3}{2} \sqrt{(V_d^* \cdot I_q^*)^2 + (V_q^* \cdot I_q^*)^2}$$  \hspace{1cm} (8)$$

It can be seen that the oscillation amplitude $P_{cs}$ and $Q_{cs}$ are equal for a given grid fault condition. For the unbalanced grid faults, the two-phase fault condition introduces larger power oscillation than the one-phase grid fault during the whole range of dip severity. The maximum power oscillation amplitude under two-phase fault is 0.5 p.u. and 0.33 p.u. for one-phase fault.

It can be expected, the significant differences in the delivered power, phase angles and modulation index under various grid fault conditions may lead to significantly different loading statuses of power semiconductor devices.

**Fig. 13.** Active/reactive power oscillation amplitude vs. Dip severity on PCC.
IV. LOSS DISTRIBUTION UNDER VARIOUS GRID FAULTS

Once the LVRT operation profiles of the grid-converter have been clarified in the previous section depending on the types and severity of voltage dips, it is possible to determine the loss distribution in the power semiconductors devices under various grid faults. The losses combined with the thermal impedance will determine the thermal behaviour of the power converter under LVRT operation, as described in the next section.

In this paper the press-pack IGCT 5SHY 40L4511 (4.5 kV/ 3.6 kA) and recommended diodes 5SDF 10H4503 are chosen as the power semiconductor devices for 3L-NPC grid connected converter [24]. The used loss model shares the same idea in [25], [26] which is a commonly accepted method for loss evaluation of power semiconductor devices, and the loss simulation is carried out based on PLECS Blockset in Simulink [26]. The average switching loss $P_{\text{sw, ave}}$ of power devices (switches or diodes) can be calculated as:
where the instantaneous switching loss $P_{sw\_inst}$ of power devices can be calculated by:

$$P_{sw\_inst}(t) = f_s \cdot E_{sw}(i_x(t), T_j) \cdot (V_{dc} / 2V_{ref})^{K_v} \tag{10}$$

$K_v$ is the voltage coefficient which can be found in [27], $V_{ref}$ is the commutated voltage tested on datasheet, $f_s$ is the switching frequency. The switching energy loss of $E_{sw}$ ($E_{on}$, $E_{off}$ for switches and $E_{sr}$ for diodes) which is function of load current $i_x(t)$ and junction temperature $T_j$ can be expressed as:

$$E_{sw}(i_x(t), T_j) = E_{sw}(i_x(t)) \cdot \left[1 + K_T \cdot (T_j - T_{ref})\right] \tag{11}$$

where the curve for $E_{sw}(i_x(t), T_j)$ can be found in the device datasheet, $K_T$ is the temperature coefficient which can be found in [27], $T_{ref}$ is the reference temperature under test in the datasheet, normally at 25 °C or 125 °C.

Similarly the average condition loss $P_{cond\_ave}$ of power devices can be calculated as:

$$P_{cond\_ave} = f_o \cdot \int_0^{1/f_o} P_{cond\_inst}(t)dt \tag{12}$$

where the instantaneous conduction loss $P_{cond\_inst}$ can be calculated as:

$$P_{cond\_inst}(t) = v_{cond}(i_x(t), T_j) \cdot i_x(t) \cdot D(m_x, t) \tag{13}$$

The duty ratio $D(m_x, t)$ is a function of the modulation index $m_x$ and related to the modulation strategy. The conduction voltage curve $v_{cond}(i_x(t), T_j)$ of power device can be found in the device datasheet, and can be analytically represented by:

$$v_{cond}(i_x(t), T_j) = \left[V_{cond0(Tref)} + K_{i2} \cdot (T_j - T_{ref})\right] + i_x(t) \cdot \left[r_{cond(Tref)} + K_{i3} \cdot (T_j - T_{ref})\right] \tag{14}$$

where $K_{i2}$, $K_{i3}$, $V_{cond0(Tref)}$ and $r_{cond(Tref)}$ are coefficients which can be found in [27] and the device datasheet.

It is noted that during the LVRT operation, the DC bus of the power converter may probably increase because of the short term mismatch in the input and output active power through the converter [8], [9]. The increased DC bus voltage should be limited (e.g. maximum 110 % rated) for hundreds of milliseconds by triggering the braking chopper, which normally consists of a resistor in series of a switch and is paralleled with the DC bus, as also indicated in Fig. 6. According to the loss model in [25], [26], the DC bus voltage has important impact on both of the switching loss and the conduction loss (modulation index) in power switching devices. As a result, the increased DC bus voltage should be taken into account in the loss analysis during LVRT operation. Moreover, the increased DC bus voltage may decrease the lifetime of power switching devices due to the cosmic radiation failure mechanism as reported in [28]. However, this issue will not be discussed in this paper.

The loss distribution of the power switching devices under the most stressed normal operation ($v_w=12$ m/s), and under the most stressed LVRT operation ($D_1=0$ p.u.), are compared in Fig. 14, where three types of grid faults are indicated in Fig. 14 (a), Fig. 14 (b) and Fig. 14 (c) respectively. The 10% higher DC bus voltage is applied for the LVRT conditions. It can be seen that, all of the three extreme LVRT operations impose the diodes and inner switches of 3L-NPC converter with significant larger losses than the normal operation condition. Moreover, the loss distribution among the three phases of converter is asymmetrical under the one-phase and two-phase unbalanced grid faults.
V. THERMAL BEHAVIOUR UNDER VARIOUS GRID FAULTS

The thermal performance of power devices are closely related to the reliability of the converter, current rating of power devices and cost of the cooling system. Therefore it is an important indicator for full-scale wind power converters. In order to conduct thermal performance evaluation, an appropriate thermal model should first be acquired.

The thermal models for a single switch and clamping diode are indicated in Fig. 15 [24], [29], in which the thermal impedance from junction to case $Z_{j-c}$ is modeled as a multi-layer Foster RC network, as shown in Fig. 16. Each of the thermal parameters can be found from the manufacturer datasheets, where the thermal resistance $R_{th}$ will decide the steady state mean value of the junction temperature, and the thermal capacitance (with time constant $\tau$) will decide the dynamic change or fluctuation of the junction temperature.

It is noted that normally the IGCT manufacturer will only provide thermal parameters inside IGCT press-pack with Foster RC network. In order to establish the complete thermal models from junction to the ambient, the thermal impedance of $Z_{T/D(j-c)}$ has to be transferred to the equivalent Cauer RC network in the simulation to facilitate the thermal impedance extension [28]. Because the temperature of the heat sink $T_h$ is normally much lower and more stable compared to the junction temperature $T_j$ in a properly designed converter system, the heat sink temperature is considered as a constant value at 60 °C in this paper. However, the heat sink temperature may strongly depend on the operation site and the design of the heat sink system.

Based on the previous loss results and thermal model from datasheet, the junction temperature of the power devices under various grid faults can be investigated by the PLECS blockset in Simulink.

A. Single-phase grounded fault

The junction temperatures for the three phases of 3L-NPC converter undergoing the extreme one-phase grounded grid fault (Type C voltage dip on Bus 2 with $D_i=0$ p.u., $v_w=12$ m/s) are shown in Fig. 17. It can be seen that the thermal loading in the three phases of the converter are slightly different from each other under this fault condition. Phase B has more stressed $D_{npc}$ and $T_{out}$ while phase A and phase C have more stressed $T_{in}$, $D_{out}$ and $D_{in}$.

The mean junction temperature $T_{jm}$ of the switches and diodes under different dip severities of one-phase grid fault are shown in Fig. 20 (a) and Fig. 21 (a) respectively. The $T_{out}$ and $D_{npc}$ are the most stressed devices within the whole dipping range. It is noted that the difference of junction temperature among three phases is not so large (around 10-15 °C max when $D_i=0$ p.u.) and only becomes significant when the dip severity $D_i$ is below 0.5 p.u. (i.e. when the converter needs to provide 100% reactive current to the power grid).

B. Two-phase connected fault

The junction temperatures for the three phases of 3L-NPC converter undergoing the extreme two-phase connected grid fault (Type D voltage dip on Bus 2 with $D_i=0$ p.u., $v_w=12$ m/s) are shown in Fig. 18. It can be seen that the thermal loading behaviors in the three phases of the converter are totally different from each other.
The mean junction temperature $T_m$ of the switches and diodes of two-phase grid fault are shown in Fig. 20 (b) and Fig. 21 (b) respectively. Compared to the single-phase grid fault condition in Fig. 20 (a) and Fig. 21 (a), the junction temperature difference among the three phases of the converter is much larger (around 20-25 °C max when $D_1=0$ p.u.). Similarly, the thermal difference among three phases of converter only becomes significant when the dip severity $D_1$ is below 0.5 p.u..

C. Three-phase grounded fault

The junction temperatures for the three phases of 3L-NPC converter undergoing the extreme three-phase grounded grid fault (Type A voltage dip on Bus 2 with $D_1\approx 0$ p.u., $v_w=12$ m/s) are shown in Fig. 19. It can be seen that the thermal loading behaviors in the three phases of the 3L-NPC converter are exactly the same under this grid fault. $D_{npc}$ and $T_{in}$ are the most stressed devices.
The mean junction temperature $T_{in}$ of the switches and the diodes under different dip severities of three-phase balance grid fault are shown in Fig. 20 (c) and Fig. 21 (c) respectively. It is interesting to see the change of junction temperature with relation to the dip severity are different when $D_1$ is below 0.5 p.u. and above 0.5 p.u.. This is because that there is 100% reactive current required by grid devices and heat sink system for the wind power converter.

D. Comparison

Fig. 22 presents the junction temperature comparison under different grid fault conditions, the $D_{npc}$, $T_{in}$ and $D_{out}$ in the most stressed phase of each grid fault are shown respectively. Special attention should be given to the power device $T_{in}$, $D_{out}$, $D_{in}$ and $D_{npc}$ under the LVRT operation of 3L-NPC. In fact those devices may have even higher junction temperature than the case of the normal operation. (Up to 40°C higher for $D_{out}$, 20°C higher for $T_{in}$, 15°C higher for $D_{in}$, and 10°C higher for $D_{npc}$). This overloading should be taken into account when designing the power devices and heat sink system for the wind power converter.

VI. CONCLUSION

Because the loading conditions of power devices in the wind turbine system under grid faults are still not yet clarified, the paper aim was to investigate the impacts of various grid voltage dips on the operating and loading behaviors of full-scale wind power converter. The analyzing methods and results are important for understanding and improving the thermal behaviors of wind power converter under this adverse condition [14], [30], [31].

It has been found that the voltage dips may vary significantly on different locations (buses) in the power grid system. Depending on the types and severity of grid faults as well as LVRT control strategies, the operation status of grid connected power converter like delivered power, phase angles and modulation index are significantly different. It is interesting to notice that the operation conditions of the power converter are unsymmetrical among the three phases when unbalanced grid faults are present.

The dramatically modified operation status under various grid faults will lead to uneven device loading not only among the three phases, but also among different types and severities of voltage dips. It should be noted that some power devices under the LVRT operation may have even higher junction temperature than the normal operation condition. This overloading should be taken into account when designing the power devices and heat sink system for the wind power converter.

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