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Wei, Baoze; Guo, Xiaoqiang; Guerrero, Josep M.; Savaghebi, Mehdi

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Leakage Current Suppression with A Novel Six-Switch Photovoltaic Grid-Connected Inverter

Baoze Wei, Student Member, IEEE, Xiaoqiang Guo, Senior Member, IEEE, Josep M.Guerrero, Fellow, IEEE, and Mehdi Savaghebi, Member, IEEE

Abstract —In order to solve the problem of the leakage current in non-isolated photovoltaic (PV) systems, a novel six-switch topology and control strategy are proposed in this paper. The inductor-bypass strategy solves the common-mode voltage limitation of the conventional six-switch topology in case of unmatched inductances. And the stray capacitor voltage of the non-isolated photovoltaic system is free of high frequency ripples. Theoretical analysis and simulation are carried out to verify the proposed topology and its control strategy. Results indicate that the leakage current suppression can be achieved with the proposed solution.

Index Terms —common-mode voltage, leakage current, grid-connected inverter, non-isolated, photovoltaic power generation

I. INTRODUCTION

ON-isolated photovoltaic power generation system has obtained more and more attention due to its advantages of small volume, low cost and high efficiency [1]. But, a common-mode current (leakage current) will be generated without the transformer. It will flow through the loop which is constituted by stray capacitance, the inverter, the filter and the grid [2-3]. It might cause grid current distortion, electromagnetic interference, and even threaten the safety of people [4-5]. So, measures should be taken to suppress the leakage current [6].

Reference [7] proposed an AC-bypass topology by adding auxiliary switches to the AC side of traditional single phase full bridge inverter, and reference [8] adds auxiliary switches to the DC side to make it as a DC-bypass topology. By doing so, the PV panels and the grid can be isolated during the current freewheeling time, and the leakage current can be effectively suppressed because of the

- B. Wei is with the Department of Energy Technology, Aalborg University, Aalborg, Nordjylland 9220 Denmark. (e-mail: bao@et.aau.dk).
- X. Guo is with Department of Electrical Engineering, Yanshan University, Qinhuangdao, Hebei 066004 China. (email: gxq@ysu.edu.cn)
- J. M. Guerrero is with the Department of Energy Technology, Aalborg University, Aalborg, Nordjylland 9220 Denmark (e-mail: ioz@et.aau.dk).
- M. Savaghebi He is with the Department of Energy Technology, Aalborg University, Aalborg, Nordjylland 9220 Denmark (e-mail: mes@et.aau.dk).

constant common-mode voltage with unipolar modulation method. But, the abovementioned strategies require that the filtering inductors on the grid side be equal to each other. However, in real applications, the inductances of the filtering inductors will be changing because of the working conditions or the changing of external environment and even they are different from the very beginning during the production process [9]; thus, it is difficult to make sure that the inductors are perfectly matching. So the leakage current problem in the case of unmatched inductances should gain more researches [10].

To solve this problem, a novel six-switch topology and control strategy are proposed in this paper. The inductor-bypass strategy can make sure that there are no high frequency components in the common-mode voltage, so that the leakage current can be effectively suppressed even in case of unmatched inductors. Simulation results verify the effectiveness of the proposed solution.

II. CONVENTIONAL SIX-SWITCH TOPOLOGY

Fig.1 shows the HERIC six-switch topology. C_{pv} is the stray capacitor between the PV array and the ground. L_1 and L_2 are the grid side filtering inductors. In the topology, S_5 and S_6 are the auxiliary switches to format the freewheeling path for the grid current. TABLE I shows its four working states. E, P and N reference grid voltage, positive half and negative half of grid voltage respectively [11]. And ON and OFF represent switched-on and switched-off respectively.

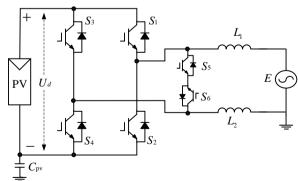


Fig.1. HERIC six-switch topology

TABLE I
SWITCHING STATES AND VOLTAGE DROP ON STRAY CAPACITOR

S_1	S_2	S_3	S_4	S_5	S_6	$U_{\it Cpv}$	States	Е
ON	OFF	OFF	ON	OFF	ON	$U_{_1}$	1	р
OFF	OFF	OFF	OFF	OFF	ON	U_2	2	r
OFF	ON	ON	OFF	ON	OFF	U_3	3	N
OFF	OFF	OFF	OFF	ON	OFF	$U_{\scriptscriptstyle 4}$	4	1N

According to TABLE I, during the positive half of grid voltage, the switches S_2 , S_3 , and S_5 remain OFF S_6 remains ON, and the switches S_1 and S_4 work with SPWM. In working state 1, switches S_1 and S_4 switched-on, the common-mode voltage U_{Cpv} equals to the voltage of the inductor L_2 , which can be expressed as (1) [7].

$$U_{1} = \frac{E - U_{d}}{L_{1} + L_{2}} L_{2} = \frac{E}{L_{1} / L_{2} + 1} - \frac{U_{d}}{L_{1} / L_{2} + 1}$$
(1)

When S_1 and S_4 switched-off, the common-mode voltage U_{Cpv} equals to the sum of the voltage of the inductor L_2 and the voltage drop on the switch S_4 , which can be expressed as (2).

$$U_2 = \frac{E}{L_1 + L_2} L_2 - \frac{U_d}{2} = \frac{E}{L_1 / L_2 + 1} - \frac{U_d}{2}$$
 (2)

With the same principle, we can get the common-mode voltages on the conditions of other two working states during the negative half of the grid voltage as (3) and (4).

$$U_{3} = E - \frac{E + U_{d}}{L_{1} + L_{2}} L_{1} = \frac{EL_{2} / L_{1} - U_{d}}{1 + L_{2} / L_{1}}$$
(3)

$$U_4 = \frac{E}{L_1 + L_2} L_2 - \frac{U_d}{2} = \frac{E}{L_1 / L_2 + 1} - \frac{U_d}{2}$$
 (4)

So, in the case of $L_1=L_2$, the common-mode voltage will maintain in constant $(E-U_{\rm d})/2$. It only contains DC component $-U_{\rm d}/2$ and low frequency component E/2. According to the equation $i=C_{\rm pv}(dU_{\rm Cpv}/dt)$, the leakage current can be effectively suppressed [7]. But, if the filtering conductors are not perfectly matching, we can see that the common-mode voltage will changing in a high frequency according to $(1)\sim(4)$.

The same problem exists in other DC-bypass or AC-bypass topologies [6], so more attention should be paid on suppressing the leakage current in the case of unmatched inductors [10].

III. PROPOSED SIX-SWITCH TOPOLOGY

In order to solve the abovementioned problem, a novel six-switch inductor-bypass topology is proposed as shown in Fig. 2. The IGBTs S_1 , S_2 , S_3 and S_4 are the main switches, and the reverse-blocking (RB) IGBTs S_5 and S_6 are the auxiliary switches. With the RB-IGBT, it will reduce the switching loss compared to the combination of a normal IGBT and a diode[12].

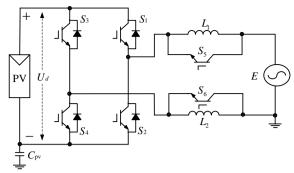
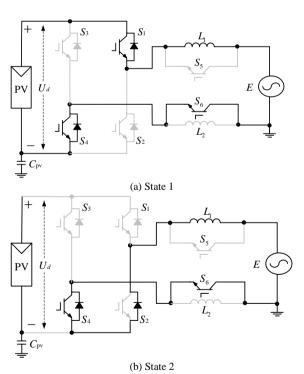


Fig. 2. Proposed six-switch topology

Compared to the HERIC topology, the filtering inductors L_1 and L_2 are paralleling connected with RB-IGBTs S_5 and S_6 respectively. With the proposed control strategy, the inductor L_1 works in the positive half of the grid voltage and the inductor L_2 works in another half. In real application, L_1 and L_2 can be replaced by a coupling inductor to increase the power density [13].

The working states and the modulation strategy are shown in Figs. 3 and 4, respectively. TABLE II shows the switching states and the corresponding common-mode voltage.

During the positive half of grid voltage, switches S_4 and S_6 remain ON, and switches S_2 , S_3 , S_5 remain OFF, only S_1 works in high frequency with SPWM. According to Figs. 3(a) and (b), the common-mode voltage U_{Cpv} will be 0 whenever switch S_1 is in ON or OFF state.



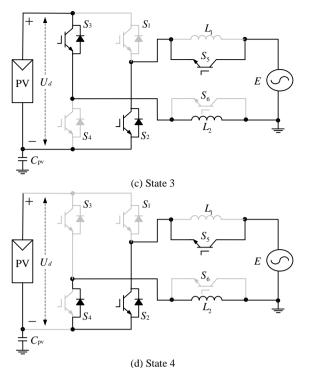


Fig. 3. Operation modes of proposed topology

During the negative half of grid voltage, switches S_2 and S_5 will remain ON, and switches S_1 , S_4 and S_6 will remain OFF, only switch S_3 works in high frequency with SPWM. From Fig. 3(c) and (d), we notice that the common-mode voltage U_{Cpv} will be the grid voltage $(U_m sin(\omega t + \psi))$ whenever switch S_3 is in ON or OFF state. And because the frequency of the grid voltage is much lower than the switching frequency, so it will just have a small effect to the leakage current.

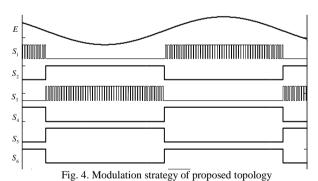


TABLE II SWITCHING STATES AND VOLTAGE DROP ON STRAY CAPACITOR

 U_{Cpv} S_1 S_2 S_3 S_4 S_5 S_6 Е OFF OFF ON OFF ON ON 0 P OFF ON ON 0 OFF OFF OFF OFF ON OFF OFF ON $U_{...}\sin(\omega t + \varphi)$ N OFF ON OFF OFF ON OFF $U_m \sin(\omega t + \varphi)$

IV. SIMULATION RESULTS

The control strategy of the novel six-switch topology is shown in Fig. 5. The current hysteresis control is used in the MATLAB simulation.

The principle of the control strategy is expressed as follows: according to TABLE II, only switch S1 works in high frequency during the positive half of grid voltage (E>0). When the error of grid current i_g and the reference current ig^* is higher than the set point Δi $(i_g>i_g^*+\Delta i)$, S_1 is switched-off; when the error is lower than $-\Delta i$ $(i_g< i_g^*-\Delta i)$, S_1 is switched-on; and if grid current ig matches $(i_g^*-\Delta i)$ is switched-on; and if grid current ig matches $(i_g^*-\Delta i)$, S_1 remains the former state, as expressed in (5). During the negative half, the control logic is shown in (6).

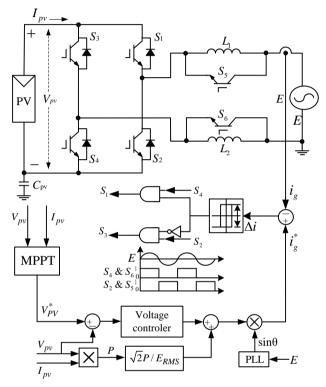


Fig. 5. System control strategy

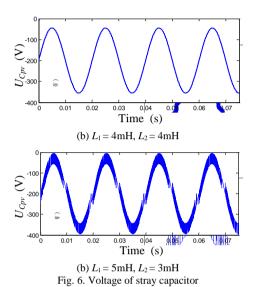
$$E > 0 \begin{cases} i_{g}(t) > i_{g}^{*}(t) + \Delta i & S_{1} & \text{OFF} \\ i_{g}(t) < i_{g}^{*}(t) - \Delta i & S_{1} & \text{ON} \\ S_{4}, S_{6} & \text{ON}, S_{2}, S_{3}, S_{5} & \text{OFF} \end{cases}$$
(5)

$$E < 0 \begin{cases} i_{g}(t) > i_{g}^{*}(t) + \Delta i & S_{3} & \text{OFF} \\ i_{g}(t) < i_{g}^{*}(t) - \Delta i & S_{3} & \text{ON} \\ S_{4}, S_{6} & \text{OFF}, S_{1}, S_{2}, S_{5} & \text{ON} \end{cases}$$
(6)

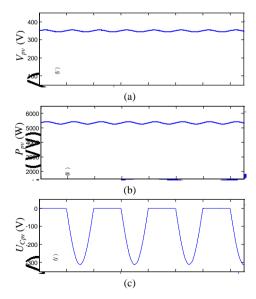
In order to verify the effectiveness of the proposed novel six-switch topology and its control strategy in the case of unmatched inductors, in the simulation the inductances of the filtering inductors are different from each other. The parameters of the simulation are shown in TABLE III.

TABLE III
PARAMETERS OF THE SIMULATION

Maximum power of PV array	5.4kW		
Voltage at P _{max} (V _{mp})	356V		
Current at P _{max} (I _{mp})	15.15A		
stray capacitance	470nF		
filtoning industrans	$L_1=5\text{mH}$		
filtering inductors	$L_2=3\text{mH}$		
Light intensity	1000W/m^2		
Temperature of PV panels	25°C		



Simulation results of the HERIC are shown in Fig. 6 with both cases of matched and unmatched filtering inductors. It can be seen that the common-mode voltage U_{Cpv} contains both low frequency and high frequency in the case of unmatched filtering inductors which is the same as analysis.



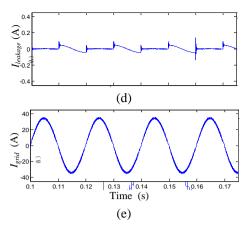


Fig. 7. Simulation results of the proposed topology. (a) Voltage of the PV array. (b) Out-put power of the PV array. (c) Common-mode voltage. (d) Leakage current. (e) Grid current.

Fig. 7 shows the simulation results of the proposed six-switch topology and the corresponding control strategy in the case of unmatched filtering inductors. According to the wave forms of V_{pv} and P_{pv} , the PV array reaches the maximum power; and it can be noticed that the common-mode voltage U_{Cpv} only contains DC and low frequency components compared to the Fig. 6. The leakage current $I_{leakage}$ can be effectively suppressed, its RMS value is lower than 25mA, and its peak value is lower than 200mA. It matches the German DIN VDE 0126-1-1 standard [14]. At the same time, sinusoidal grid current I_{grid} is obtained.

V. CONCLUSION

This paper has proposed a novel six-switch inductorbypass topology and a corresponding control strategy. With this solution, the leakage current can be effectively suppressed even in the case of unmatched filtering inductors which is different from the existing AC-bypass or DCbypass topologies. And there is only one switch working in high frequency at any time which can reduce the switching loss.

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Baoze Wei received the B.S. degree in electrical engineering, and the M.S. degree in power electronics and power drives from the Yanshan University, Qinhuangdao, China, in 2010 and 2014, respectively. He is currently working toward the Ph.D. degree in power electronic systems, from the Department of Energy Technology, Aalborg University, Aalborg, Denmark.

His research interests include microgrid, modular power inverters, photovoltaic generation system, paralleling power converter for renewable generation systems, power quality, as well as the applications of distributed control.

Xiaoqiang Guo (M' 10-SM' 14) received the B.S. and Ph.D. degrees in electrical engineering from Yanshan University, Qinhuangdao, China, in 2003 and 2009, respectively.

He has been a Postdoctoral Fellow with the Laboratory for Electrical Applications and Research (LEDAR), Ryerson University, Toronto, ON, Canada. He is currently an associate professor with the Department of Electrical Engineering, Yanshan University, China. He has authored/coauthored more than fifty technical papers, in addition to nine patents. His current research interests include high-power converters and ac drives, electric vehicle charging station, and renewable energy power conversion systems.

Dr. Guo is a Senior Member of the IEEE Power Electronics Society and IEEE Industrial Electronics Society. He is an active Referee for IEEE Transactions on Industrial Electronics and IEEE Transactions on Power Electronics.

Josep M. Guerrero (S'01–M'04–SM'08–F'14) received the B.S. degree in telecommunications engineering, the M.S. degree in electronics engineering, and the Ph.D. degree in power electronics from the Technical University of Catalonia, Barcelona, Spain, in 1997, 2000, and 2003, respectively.

Since 2011, he has been a Full Professor with the Department of Energy Technology, Aalborg University, Aalborg, Denmark, where he is responsible for the Microgrid Research Program. From 2012, he has been a Guest Professor at the Chinese Academy of Science, Beijing, China, and

the Nanjing University of Aeronautics and Astronautics, Nanjing, China, and from 2014, he has been a Chair Professor at Shandong University, Shandong, China. His research interests include oriented to different microgrid aspects, including power electronics, distributed energy-storage systems, hierarchical and cooperative control, energy management systems, and optimization of microgrids and islanded minigrids.

Prof.Guerrero is an Associate Editor for the IEEE TRANSACTIONS ON POWER ELECTRONICS, the IEEE TRANSACTIONS ON INDUSTRIAL ELECTRONICS, and the IEEE INDUSTRIAL ELECTRONICSMAGAZINE, and an Editor for the IEEE TRANSACTIONS ON SMART GRID. He has been a Guest Editor of the following IEEE TRANSACTIONS ON POWER ELECTRONICS, special issues on Power Electronics for Wind Energy Conversion and Power Electronics for Microgrids; the IEEE TRANSACTIONS ON INDUSTRIAL ELECTRONICS special sections on Uninterruptible Power Supplies Systems, Renewable Energy Systems, Distributed Generation and Microgrids, and Industrial Applications and Implementation Issues of the Kalman Filter; and the IEEE TRANSACTIONS ON SMART GRID special issue on Smart DC Distribution Systems. He was the Chair of the Renewable Energy Systems Technical Committee of the IEEE Industrial Electronics Society. In 2014, he received the ISI Highly Cited Researcher by Thomson Reuters.

Mehdi Savaghebi was born in Karaj, Iran, in 1983.He received the B.Sc. degree from University of Tehran, Iran, in 2004, and the M.Sc. (Hons.) and Ph.D. (Hons.) degrees from the Iran University of Science and Technology, Tehran, in 2006 and 2012, respectively, all in electrical engineering.

From 2007 to 2014, he was a Lecturer in the Electrical EngineeringDepartment, KarajBranch, Islamic Azad University, Karaj, Iran, where he taught various courses and conducted research on power systems and electrical machines. In 2010, he was a Visiting Ph.D. Student with the Department of Energy Technology, Aalborg University, Aalborg, Denmark, and with the Department of Automatic Control Systems and Computer Engineering, Technical University of Catalonia, Barcelona, Spain. He is currently a Postdoctoral Research Assistant in the Department of Energy Technology, Aalborg University. His main research interests include distributed generation systems, microgrids, and power quality issues of electrical systems.