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Published in:

DOI (link to publication from Publisher):
10.1109/APEC.2014.6803600

Publication date:
2014

Document Version
Early version, also known as pre-print

Link to publication from Aalborg University

Citation for published version (APA):

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Autonomous Active and Reactive Power Distribution Strategy in Islanded Microgrids

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Abstract — This paper proposes an autonomous active and reactive power distribution strategy that can be applied directly on current control mode (CCM) inverters, being compatible as well with conventional droop-controlled voltage control mode (VCM) converters. In a microgrid, since renewable energy sources (RES) units regulate different active power, the proposed reactive power distribution is adaptively controlled according to the active power distribution among energy storage systems (ESS) and RES units. The virtual impedance is implemented in order to improve the reactive power sharing in a distributed way. Real-time hardware-in-the-loop results are presented to verify the proposed control strategy.

I. INTRODUCTION

Microgrids are providing a promising way to integrate distributed generators (DG) such as renewable energy sources (RES) and energy storage systems (ESS) to supply power to local loads [1]. Among the control strategies investigated in islanded microgrids, master-slave control structure is the most popular way to manage the ESS and RES units, where the ESS units operate as grid forming units and RES units operate as grid following units [2].

Usually, RES units control output active power by using maximum power point tracking strategies, and deliver no reactive power. Therefore almost all of the reactive power from demand side is provided by the ESS units. However, considering the total output power rating of the converter is fixed, the more reactive power delivered from ESS units, the less capacity of active power regulation of ESS used to buffer the unbalance of power generation and consumption can be provided. In order to overcome such a limitation, many previous researches have been carried out to investigate the implementation that using RES units to provide reactive power support and improve power quality [3], [4]. However, these methods are based on setting fixed reactive power commands of RES units intensively. And the proportional active and reactive power sharing should be fulfilled by using a microgrid central controller to adjust the power references of RES units with additional communication link [5].

Alternatively, in order to achieve the power sharing in a fully decentralized way, droop method can be a good solution for the voltage controlled mode (VCM) converters [6], [7]. A frequency-bus-signaling control for the power management based on droop control is proposed in [8], and a coordinated control for the active power distribution among ESS and RES units is proposed in [9]. However, the conventional reactive power-to-voltage (Q–V) droop method is mostly used on VCM inverters, which is difficult to be applied directly on RES inverters since they are often current controlled mode (CCM) inverters with only current inner loop. In this sense, this paper presents a voltage-to-reactive power (V–Q) reverse droop control strategy that can be applied on CCM inverters to allow VCM/CCM units coordination when sharing the demanded reactive power. In order to ensure proper autonomous power sharing based on basic droop and reverse droop method, this paper proposes an advanced primary control that taking into account of the system capacity. Furthermore, since the voltage drop over the line impedance will create inaccuracy of reactive power distribution, this phenomenon is also discussed in this paper.

This paper is organized as follows. The principle of the basic primary control, and advanced primary control taking into account of the system capacity are illustrated in the Section II. Section III provides the analysis for the accuracy of reactive power sharing and the corresponding solutions for the improvement. Section IV describes the system control implementation. The real-time hardware-in-the-loop results are presented in Section V. Finally Section VI gives the conclusion.

II. PRIMARY CONTROL FOR VCM AND CCM

A. Droop and Reverse Droop Control

An islanded microgrid with both VCM and CCM units is shown in Fig. 1. In order to achieve autonomous active/reactive power sharing among DG units operating in VCM, droop control is often used in primary loop (considering output impedance is inductive) [6].
This document is a pre-print of the final paper:

The droop control characteristic of active/reactive power regulation is shown in Fig. 2.

For the DG units operating in CCM, the active and reactive power is regulated directly by inner current loop. In order to contribute to regulate the AC bus frequency/amplitude, reverse droop can be applied as,

\[
P_c = \frac{1}{m_r} (\omega^* - \omega_e)
\]

(3)

\[
Q_c = \frac{1}{n_r} (E^* - E_e)
\]

(4)

where \(\omega_e\) and \(E_e\) are the measured grid frequency and voltage of CCM units, \(P_c\) and \(Q_c\) are the output active and reactive power of CCM converters, \(m_r\) and \(n_r\) are the droop coefficients. The reverse droop control characteristic of frequency/voltage regulation is shown in Fig. 3.

Supposing \(E=E_e\) and \(\omega=\omega_e\) in ideal case of measurement, and ignoring the inner-loop regulation with higher bandwidth, we have the power distribution of the DG system integrating both VCM and CCM units by combing (1)-(4),

\[
P_1 : P_2 : \cdots : P_i = \frac{1}{m_1} : \frac{1}{m_2} : \cdots : \frac{1}{m_i}
\]

(5)

\[
Q_1 : Q_2 : \cdots : Q_i = \frac{1}{n_1} : \frac{1}{n_2} : \cdots : \frac{1}{n_i}
\]

(6)

where \(P_i\) and \(Q_i\) are the output active and reactive power of each unit, \(m_i\) and \(n_i\) are the droop/reverse droop coefficients of active and reactive power respectively. According to (5) and (6), the active/reactive power distribution can be simply achieved by assigning proper sets of coefficients of \(m_i\) and \(n_i\) in a distributed way.

**B. Primary Control Considering System Capacity and Power Availability**

In practical cases, the active/reactive power distribution should take into consideration of power availability of RES operating in CCM, and capacity constraint of each system. In this sense, the active power regulation for VCM and CCM should meet the demand of (i) the active power of CCM operates at maximum power point (MPP) and (ii) the VCM units share the remained active power according to the capacity. Therefore, the active power command of CCM shown in (3) is changed as

\[
P_c = P^*
\]

(7)

where \(P^*\) denotes the generated maximum power reference based on the energy sources condition. And the droop coefficient of VCM is designed as

\[
0 < m_d \leq \frac{\Delta\omega}{P_{\text{max}V}}
\]

(8)

where \(P_{\text{max}V}\) and \(\Delta\omega\) are maximum output active power and corresponding bus frequency deviation respectively.

In terms of the reactive power regulation, the maximum available reactive power of each unit is expressed as

\[
Q_{\text{max}} = \sqrt{S_{\text{max}}^2 - P^2}
\]

(9)

where \(Q_{\text{max}}\) and \(S_{\text{max}}\) are the maximum reactive power and apparent power of each unit, \(P\) is instantaneous power. It is
shown that the reactive power availability is not only related to the power rating of the unit but also the instant output active power. The relationship of power regulation of two units is shown in Fig. 4, which shows that the more active power the unit supplies, the less capacity of reactive power regulation is remained under the same total apparent power.

Therefore, instead of independently controlling the reactive power with constant coefficient \( n_d \) and \( n_i \) for (2) and (4), autonomous reactive power control can be achieved with adaptively adjusting the droop/reverse droop coefficients \( n \) as

\[
n = \frac{\Delta E}{S_{\text{max}}^2 - P^2}
\]

where \( \Delta E \) is maximum bus voltage amplitude deviation, \( n \) is droop/reverse droop coefficient. Both frequency and voltage amplitude deviation in (8) and (10) should be designed within the requirement of grid code. Then, based on (10), the reactive power distribution among integrated VCM and CCM units can be deduced as

\[
Q_1, Q_2, \ldots, Q_n = \sqrt{S_{\text{max}}^2 - P_1^2}, \sqrt{S_{\text{max}}^2 - P_2^2}, \ldots, \sqrt{S_{\text{max}}^2 - P_n^2}
\]

where \( \Delta E \) is maximum bus voltage amplitude deviation, \( n \) is droop/reverse droop coefficient. Both frequency and voltage amplitude deviation in (8) and (10) should be designed within the requirement of grid code. Then, based on (10), the reactive power distribution among integrated VCM and CCM units can be deduced as

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\]

III. IMPROVEMENT ON REACTIVE POWER SHARING

In islanded microgrid, all the DG units share the same frequency signal so that the active power can be well distributed based on droop control. However, the voltage drop over the line impedance can be different among DG units so that the reactive power is hard to achieve accurate distribution. Fig. 6 shows the equivalent circuit of two DG units connected by line impedance.

According to (2) and (4), the reactive power difference between DG\(_1\) and DG\(_2\) can be expressed as

\[
Q_{\text{dif}} = \frac{E_1^* - E_2^* - E_1 - E_2}{n_2 - n_1}
\]

where \( n_1 \) and \( n_2 \) are defined as (10). Supposing the two DG units have the same reactive power capacity, then we have \( n_1=n_2=n \), (12) can be deduced as

\[
Q_{\text{dif}} = \frac{\left( E_1^* - E_1 \right) - \left( E_2^* - E_2 \right)}{E_{\text{dif}} - E_{\text{dif}}^*} \cdot \frac{n}{\Delta E}
\]

where \( E_{\text{dif}} \) is the voltage drop on the line impedance, and \( E_{\text{dif}}^* = E_2 - E_1 \). It shows that the suppression of reactive power difference can be obtained by increasing \( \Delta E \). In this case, the performance of \( Q_{\text{dif}} \) deduction is constrained by the maximum voltage amplitude deviation (i.e. 10% of nominal voltage deviation according to EN 50160) and the output voltage regulation is thereby deteriorated. In the other case, the reduction of \( Q_{\text{dif}} \) can be achieved by decreasing \( E_{\text{dif}} - E_{\text{dif}}^* \). Since \( E_{\text{dif}} \) is generated by the line impedance, a secondary
control can be utilized to compensate $E_{df}$ by adjustment of $E'_{df}$ with communication [10] or using virtual impedance to compensate the voltage drop in a distributed way [11].

The principle of reactive power compensation using virtual impedance is shown in Fig. 7. Since the reactive power difference of DG units is produced by the line impedance, the idea of virtual impedance is adding $Z_v$ on the reference in the control loop to compensate the voltage drop. Supposing the line impedance is inductive, and then the virtual impedance can be designed as capacitive for the compensation. To improve the system damping, a positive virtual resistance can be added on the system. Therefore, the compensating voltage reference of DG units can be expressed as

$$
\tilde{V}_{co} = \tilde{I}_o \cdot Z_v = \tilde{I}_o \cdot (R_v - j \omega L_v)
$$

where $L_v$ and $R_v$ are the virtual inductance and resistance, $\tilde{V}_{co}$ and $\tilde{I}_o$ are the vector of compensating voltage reference and output current.

**IV. CONTROL IMPLEMENTATION**

The control implementation of DG units controlled in VCM and CCM is shown in Fig. 8, which consists of inner-loop control, primary control and virtual impedance.

A. Inner-loop

The inner-loop control of VCM and CCM targets at good tracking performance of output voltages and currents respectively. For both VCM and CCM units, typical proportional integral (PI) controller is used with reference frame transformation $T_{abc/dfq}$ and $T_{eq/abc}$, between $d$-$q$ and $abc$ reference frame. The design procedure of inner-loop control can be referred to [12]. Compared to the classical independent inner-loop control, it worth noticing that the frequency signal used in the reference frame transformation for VCM is generated from primary droop control.

B. Primary Loop

For primary loop of VCM units, droop control based on (1), (2) and (10) is used. While for CCM units, reverse droop based on (3), (4) and (10) is used. When the active power of the converter reaches the maximum value $S_{max}$, the output reactive power should be limited to zero. The low pass filter (LPF) used in the primary control aims at limiting the loop bandwidth, so that the primary control can be separately designed and the innerloop of VCM and CCM can be considered as ideal voltage and current source respectively. The voltage reference generator of VCM is presented as

$$
\nu_{ref} = E \cdot \cos \theta
$$

$$
\nu_{ref} = E \cdot \sin \theta
$$

where $\nu_{ref}$ and $\nu_{ref}$ are the voltage references sent to the innerloop. $\theta$ is the phase angle. And the current reference generator block is expressed as

$$
i_{ref} = \frac{2}{3} \left( \frac{V_{ref}}{E_g} \cdot P_C + \frac{V_{ref}}{E_g} \cdot Q_C \right)
$$

$$
i_{ref} = \frac{2}{3} \left( \frac{V_{ref}}{E_g} \cdot P_C - \frac{V_{ref}}{E_g} \cdot Q_C \right)
$$

where $i_{ref}$ and $i_{ref}$ are the current references sent to the innerloop.
TABLE I. POWER STAGE AND CONTROLLER PARAMETERS

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Value</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Power Stage</strong></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Nominal Bus Voltage</td>
<td>V</td>
<td>220</td>
<td>V</td>
</tr>
<tr>
<td>Nominal Bus Frequency</td>
<td>f</td>
<td>50</td>
<td>Hz</td>
</tr>
<tr>
<td>Filter Inductance of DG&lt;sub&gt;i&lt;/sub&gt;</td>
<td>L&lt;sub&gt;i&lt;/sub&gt;</td>
<td>1.8</td>
<td>mH</td>
</tr>
<tr>
<td>Filter Inductance of DG&lt;sub&gt;1&lt;/sub&gt; and DG&lt;sub&gt;2&lt;/sub&gt;</td>
<td>L&lt;sub&gt;o&lt;/sub&gt;</td>
<td>3.6</td>
<td>mH</td>
</tr>
<tr>
<td>Filter Capacitance</td>
<td>C</td>
<td>9</td>
<td>µF</td>
</tr>
<tr>
<td>Line Inductance</td>
<td>L&lt;sub&gt;r&lt;/sub&gt;</td>
<td>1.8</td>
<td>mH</td>
</tr>
<tr>
<td>Line Resistance</td>
<td>r&lt;sub&gt;r&lt;/sub&gt;</td>
<td>0.1</td>
<td>Ω</td>
</tr>
<tr>
<td>Load</td>
<td>R&lt;sub&gt;L&lt;/sub&gt;</td>
<td>96.8, 0.38</td>
<td>Ω/H</td>
</tr>
</tbody>
</table>

**Innerloop Control**

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Value</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Voltage Loop PI</td>
<td>k&lt;sub&gt;vp&lt;/sub&gt;, k&lt;sub&gt;vq&lt;/sub&gt;</td>
<td>0.1, 200</td>
<td>- , s&lt;sup&gt;-1&lt;/sup&gt;</td>
</tr>
<tr>
<td>Current Loop PI</td>
<td>k&lt;sub&gt;ip&lt;/sub&gt;, k&lt;sub&gt;iq&lt;/sub&gt;</td>
<td>15, 50</td>
<td>- , s&lt;sup&gt;-1&lt;/sup&gt;</td>
</tr>
</tbody>
</table>

**Primary Control/Virtual Impedance**

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Value</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Frequency Droop</td>
<td>m&lt;sub&gt;d&lt;/sub&gt;</td>
<td>0.0005</td>
<td>rad s&lt;sup&gt;-1&lt;/sup&gt;/W</td>
</tr>
<tr>
<td>Voltage Droop</td>
<td>n&lt;sub&gt;d&lt;/sub&gt;</td>
<td>0.0013, 0.0053</td>
<td>V/Var</td>
</tr>
<tr>
<td>Frequency Reverse Droop</td>
<td>m&lt;sub&gt;r&lt;/sub&gt;</td>
<td>0.0005</td>
<td>rad s&lt;sup&gt;-1&lt;/sup&gt;/W</td>
</tr>
<tr>
<td>Voltage Reverse Droop</td>
<td>n&lt;sub&gt;r&lt;/sub&gt;</td>
<td>0.0013, 0.0053</td>
<td>V/Var</td>
</tr>
<tr>
<td>Capacitive Virtual Impedance</td>
<td>-L&lt;sub&gt;o&lt;/sub&gt;</td>
<td>-1.75</td>
<td>mH</td>
</tr>
</tbody>
</table>

C. Virtual Impedance

Capacitive virtual impedance is utilized to compensate the reactive power difference. The virtual impedance should be designed near to the value of line impedance to cancel the voltage drop. For d-q reference frame control system, (14) is rewritten as

\[
\begin{align}
    v_{vd} &= R_c \cdot i_{vd} + \omega L_c \cdot i_{vq} \\
    v_{vq} &= R_c \cdot i_{vq} - \omega L_c \cdot i_{vd}
\end{align}
\]

(19) (20)

where \(v_{vd}, v_{vq}\) and \(i_{vd}, i_{vq}\) represent the d-q components of compensating voltages and output currents. Then the compensating voltage is added to the voltage reference in the innerloop to decrease \(E_{diff}\).

V. HARDWARE-IN-THE-LOOP RESULTS

Hardware-in-the-loop simulations are carried out based on dspace1006 platform to validate autonomous active and reactive power distribution control strategy for VCM and CCM units. The system configuration is shown in Fig. 9, where the DG<sub>1</sub> operates in VCM with droop control, and DG<sub>2</sub> and DG<sub>3</sub> operate in CCM with reverse droop. The power stage and control system parameters are shown in Table I.

Fig. 10 shows the simulation results of active and reactive power distribution with the variation of droop/reverse droop coefficients. For DG<sub>1</sub>, droop control is applied on both active and reactive power, while for DG<sub>2</sub> and DG<sub>3</sub> reverse droop is only applied on reactive power and active power control utilizes a constant power tracking at 1kW. In order to evaluate the reactive power sharing accuracy, the droop/reverse droop coefficients is selected the same as \(n_d=n_r=0.0013\). DG<sub>2</sub> starts up at \(t_1\), and supply active and reactive power to the grid automatically. It can be seen reactive power difference \(Q_{diff}\) is

434Var. At \(t_2\), the droop/reverse droop coefficients increase to \(n_d=n_r=0.0053\), and \(Q_{diff}\) decreases to 140Var. It can be seen that increasing droop/reverse droop coefficients can increase
the reactive power sharing accuracy. But at the same time the bus voltage deviation deteriorates that $E$ changes from 1V to 4V.

Fig. 11 shows the simulation results of active and reactive power distribution with negative virtual impedance. DG1 starts up at $t_1$, and the negative virtual impedance is activated at $t_2$. It can be seen reactive power difference $Q_{\text{diff}}$ is effectively suppressed from 434Var to 15Var without deteriorating the bus voltage deviation. It proves that the reactive power distribution can be more accurate by using virtual impedance compared to the $Q_{\text{diff}}$ when the two units sharing equal reactive power.

Fig. 12 shows the simulation results of active and reactive power distribution with consideration of system capacity. In this case, the reactive power of three DG units is distributed adaptively to the active power regulation. DG1 starts up at $t_1$, and output active power 2kW, so that DG1 charges 0.5kW active power. According to (9), $Q_{\text{max1}}/Q_{\text{max2}}=1.32:1$. DG2 should distribute more reactive power than DG3, so that it can be seen from the results DG1 and DG3 provide 673Var and 513Var respectively. Then DG2 starts up at $t_2$, and provide 1kW active power to loads. In this scenario, $Q_{\text{max1}}/Q_{\text{max2}}/Q_{\text{max3}}=1.16:1.27:1$, so that DG1, DG2 and DG3 share the reactive power of loads at 400Var, 440Var and 348Var respectively according to power capacity of each converter at 3kVA, the small difference of reactive power distribution is due to the tolerance of virtual impedance compensation.

VI. CONCLUSION

This paper presented autonomous active and reactive power distribution for integrated CCM and VCM units in islanded microgrids. Based on the droop and reverse droop method, the proper active and reactive power management can be conveniently obtained by assigning primary loop coefficients. The droop and reverse droop control based on the system capacity was proposed so that the reactive power distribution can be adaptively controlled according to active power distribution. And the methods to reduce the reactive power sharing difference were provided. Finally, the real-time hardware-in-the-loop simulation results were given and verified the proposed control strategy.

REFERENCES


