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Maximum Power Point Tracking Implementation by Dspace Controller Integrated Through Z-Source Inverter Using Particle Swarm Optimization Technique for Photovoltaic Applications

Kalaiarasi N. 1,* [ID](https://orcid.org/0000-0002-6277-8809) , Subranshu Sekhar Dash ¹ , Sanjeevikumar Padmanaban 2,* [ID](https://orcid.org/0000-0003-3212-2750) , Paramasivam S. ³ and Pandav Kiran Morati ⁴

- ¹ Department of Electrical and Electronics Engineering, SRM University, Chennai 603203, India; sekhardash.s@ktr.srmuniv.ac.in
- ² Department of Energy Technology, Aalborg University, Esbjerg 6700, Denmark
³ Department of the Channai 600001 India: paramethy@vaboo.com
- ³ Danfoss Industries Ltd., Chennai 600001, India; paramsathya@yahoo.com
- ⁴ Department of Electrical and Electronics Engineering Science, University of Johannesburg, Auckland Park, Johannesburg 2006, South Africa; kiranpandav88@yahoo.co.in
- ***** Correspondence: kalaiarasi.n@ktr.srmuniv.ac.in (K.N.); sanjeevi_12@yahoo.co.in (S.P.); Tel.: +91-988-435-6933 (K.N.)

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Abstract: Maximum Power Point Tracking (MPPT) technique is used to extract maximum power from the photovoltaic system. This paper involves working on an enhanced Particle Swarm Optimization (PSO) based MPPT method for the photovoltaic (PV) system integrated through Z-Source inverter. The main benefit of the proposed method is the diminishing of the steady-state oscillation when the maximum power point (MPP) is located. Additionally, during an extreme environmental condition, such as partial shading and large fluctuations of irradiance and temperature, the proposed method has the capability to track the MPP. This algorithm is implemented in dspace 1104 controller. MATLAB simulations are carried out under varying irradiance and temperature conditions to evaluate its effectiveness. Its performance is compared with a conventional method like Perturb and observe (P&O) method.

Keywords: dspace controller; non-shoot through state; PSO; PV array; shoot through state; SPWM technique; Z-Source inverter

1. Introduction

Photovoltaic (PV) system consists of an array, a power conditioner(s), and a transmission and distribution system. With changing irradiation and the cell temperature, the power produced by PV module varies nonlinearly. To achieve maximum power at all times, the Maximum Power Point Tracking Technique (MPPT) is developed. Till now, the MPPT algorithms comprise the constant voltage tracking (CVT), voltage feedback, Perturb and Observe (P&O), power feedback, incremental conductance (INC) method, Hill Climbing, Artificial Neural Network (ANN), Fuzzy Logic, and so on [\[1–](#page-16-0)[6\]](#page-16-1). The simplest approach is the voltage feedback, where the output voltage at the maximum power point is essentially recognized in advance of the tracking operation. To locate the MPP, the P-V characteristic curve must be updated often, but due to the PV module declines, MPP drifts. Most of the conventional methods based on the perturbation of voltage, current or both. The popular technique is P&O method. This method compares present and past power, and then check whether the power is improved or not. If the power is improved the perturbation direction is correct otherwise the direction is reversed. This method produces more oscillation and tracking speed is less [\[7–](#page-16-2)[15\]](#page-17-0).

Over recent years, a large number of studies have been published on smart MPP trackers to accurately locate the MPP and enhance the performance of tracking, during static and dynamic conditions. In general, practically partially shaded panels have multiple peaks. But, these tracking approaches fail to track the global MPP due to the multiple peak problems [\[16](#page-17-1)-19]. In order to overcome this problem, PSO algorithm is used to follow the global MPP instead of tracking the local MPP on a multifaceted nonlinear output curve. PSO algorithm based on maximum power point tracking techniques have been proposed recently for a photovoltaic module, to increase the dynamic response of tracking $[20-28]$ $[20-28]$.

Various literature shows that the PSO based MPPT incorporates for a boost converter in PV applications [\[20](#page-17-3)-29]. The significant contribution of this work is that the PSO based MPPT is incorporated in Z-Source inverter. Also, the implementation of MPPT and PWM generation is performed in the dspace 1104 controller. This paper mentions an enhanced MPPT method based on a PSO algorithm using a Z-Source inverter. This MPPT has a comparable structure to the conventional P&O method; therefore, a direct duty cycle control can be utilized. The main advantages of the proposed method are the reduction of steady-state oscillation and faster tracking of MPP even during the extreme environmental condition, e.g., partial shading condition and large fluctuations of irradiance. By comparing it to conventional MPPT techniques, it has been found that the PSO has a faster tracking speed. Moreover, due to the simplicity of the algorithm, it can be computed very quickly; so, it is also possible to implement this algorithm using a simple low-priced controller. The tracking performance of this method is established by experimental results. In [\[30\]](#page-17-6), the algorithm computes the value of initial particles' dMPP (duty cycle at MPP) based on the voltage at maximum power. Therefore, the algorithm can start the optimization process with an initial value that is already close to the MPP.

This paper is prepared as Section [1](#page-1-0) presents introduction and necessity of MPPT, and Section 2 gives the modeling of 60 W PV panel using mathematical equations [\[31–](#page-17-7)[33\]](#page-17-8). Sections 3 and 4 discuss gives the modeling of 60 W PV panel using mathematical equations [31–33]. Sections [3](#page-4-0) and [4](#page-7-0) discuss conventional perturb and observe method and particle swarm optimization method, respectively. conventional perturb and observe method and particle swarm optimization method, respectively. To integrate the output of PV panel to load through Z-Source inverter is discussed in Section 5 [34-[38\]](#page-18-0). Finally, the results are conferred i[n S](#page-11-0)ection 6.

2. Modeling of PV Module 2. Modeling of PV Module

The PV module is used to convert light into electricity. To model the PV module, the required input variables are weather data, such as irradiance and temperature and output variables, may be voltage, current, power, or all the three. Though, to trace the I-V or P-V characteristics, there is a need for these three variables. The output changes immediately if any change in input. So, it is important to design an accurate model for the PV module [31]. Th[e PV](#page-17-7) module is used to convert light into electricity. To model the PV module, the required
input variables are weather data, such as irradiance and temperature and output variables, may be
voltage, current, power, or al

In solar panels, Maximum Power Point Tracking (MPPT) is used to achieve the maximum possible power extraction during variations of irradiance, shading, and temperature. The equivalent circuit of PV cell shown in Figure 1. In solar panels, Maximum Power Point Tracking (MPPT) is used to achieve the maximum possible power extraction during variations of irradiance, shading, and temperature. The equivalent circuit of PV cell shown [in](#page-2-1) Figure 1.

Figure 1. The equivalent circuit of photovoltaic (PV) cell.

From the PV equivalent circuit shown in Figure [1,](#page-2-1) the following parameters are calculated. The output current from the equivalent circuit is described as Equation (1) From the PV equivalent circuit shown in Figure 1, the following parameters are calculated. The

$$
I = I_{ph} - I_{sc} \left(\exp \frac{q(V + R_s I)}{NKT} - 1 \right) - \left(\frac{V + R_s I}{R_{sh}} \right)
$$
 (1)

where *I_{ph}*-photocurrent, *I*_{*sc*}-short circuit current, *V*-cell voltage, *R*_{*s*}-series resistance, and *R*_{*sh*}-shunt resistance, respectively. *Iph* is expressed as Equation (2). where *I* and *I* control control voltage, *Rs*-series resistance, and *Rs*-series resistance, and *Rsh*-shunter, and *Rsh*-shunter, and *Rsh*-shunter, and *Rsh*-shunter, and *Rsh*-shunter, and *Rsh*-shunter, and *Rsh*-shu

$$
I_{ph} = \frac{[I_{sc} + K_i(T - 298)]I_r}{1000}
$$
 (2)

where *Kⁱ* -temperature coefficient, *T*-cell temperature, and *Ir*-solar insolation in Watts/m² . The saturation diode current is specified in Equation (3). Where E_{go} -Band gap energy, Tr-Nominal temperature, V_t -Diode thermal voltage. 1000
-
Imperature, and L-solar insolation in Watts/m². The saturation

Reverse saturation diode current is

$$
I_o = I_{rs} \left[\frac{T}{T_r} \right]^3 \exp \left[\frac{E_{go}}{nV_t} \left(\frac{T}{T_r} - 1 \right) \right]. \tag{3}
$$

Solar cells are arranged in series and parallel combination to obtain a PV module. Where *Ns*-Number of cells in series and *Np*-Number of cells in parallel. In MSX 60 W panel, the number of cells connected in series is 36 and parallel are 1. The Equation (4) represents the output current of the PV module. **Property module** in the number of cells in $\frac{1}{1}$ in MSX 60 W panel, the number of cells in the number of cells in $\frac{1}{1}$ in

Output current of the PV module is 36 and parallel are 1. The Equation (4) represents the output current of the PV module is PV module.

$$
I = N_p I_{ph} - N_p I_o \left[exp \left(\frac{q \left(\frac{V}{N_s} + \frac{InR_s}{N_p} \right)}{KnT} \right) - 1 \right] - \left(\frac{\frac{N_p V}{N_s} + IR_s}{R_{sh}} \right). \tag{4}
$$

These equations are modeled in MATLAB and the P-V, I-V curves are obtained in different These equations are modeled in MATLAB and the P-V, I-V curves are obtained in different irradiance and temperatures are shown in Figures 2 and 3. Table 1 shows the parameter specifications irradiance and temperatures are shown in [F](#page-3-0)igu[re](#page-4-1)s 2 [and](#page-4-2) 3. Table 1 shows the parameter of the MSX-60Watt panel.

Figure 2. PV characteristics of MSX 60W PV panel for (a) varying irradiance (b) varying temperature.

Figure 3. I-V characteristics of MSX 60W PV panel for (a) varying irradiance (b) varying temperature.

Table 1. MSX-60W PV Module Specifications. **Table 1.** MSX-60W PV Module Specifications.

Parameter Specification	Values
Peak power, $P_{m\nu\nu}$	60 W
Peak Power voltage, V _{mpp}	17.1 V
Peak Power Current, I _{mpp}	3.5A
Short-circuit current, I_{sc}	3.8 A
Open circuit voltage, V_{oc}	21.1 V
Temperature co-efficient of current, K_i	0.003 (mA/ \degree C)
Temperature co-efficient of voltage, K_V	-0.08 (mV/ $^{\circ}$ C)
Solar irradiation at STC	1000 W/m ²
Temperature at STC	$25^{\circ}C$
Electron charge, q	1.6×10^{-19} C
Number of cells connected in series, N_s	36
Number of cells connected in parallel,	1
N_p	
Ideality factor of the diode, n	1.3
Boltzmann's constant, k	1.3805×10^{-23} J/K
Band gap energy, E_{g0}	1.12 eV
Series resistance, R_s	0.0018Ω
Shunt resistance, R_{sh}	$400\ \Omega$

3. Classification of MPPT 3. Classification of MPPT

3.1. P&O Method

change in the power of the panel. The conditions are, if $dP/dV = 0$, MPP is reached. Otherwise, dP/dV < 0, reduce the duty cycle by a value and $dP/dV > 0$, increase the duty cycle by a value [\[4–](#page-16-3)[6\]](#page-16-1). Figure 4 represents the P&O based MPPT and step by step process of the P&O algorithm is depicted in the flowchart shown in Figure [5.](#page-5-1) In perturb and observe method (P&O) a voltage perturbation is provided with respect to the

Figure 4. Block diagram of Perturb and Observe method. **Figure 4.** Block diagram of Perturb and Observe method. **Figure 4.** Block diagram of Perturb and Observe method.

Figure 5. Flowchart of Perturb and Observe method. **Figure 5.** Flowchart of Perturb and Observe method.

3.2. Hill Climbing Method 3.2. Hill Climbing Method

Hill Climbing (HC) method follows the same procedure as like of P&O method. The key difference Hill Climbing (HC) method follows the same procedure as like of P&O method. The key between both methods is, in HC method perturb is used as duty ratio for power converter and in P&O method the working voltage of the PV array [is p](#page-18-1)erturbed [39].

3.3. Incremental Conductance Method 3.3. Incremental Conductance Method

The Incremental Conductance (INC) method is faster when compared to the HC and P&O method. The Incremental Conductance (INC) method is faster when compared to the HC and P&O The logic of INC method is at MPP, the derivative of power of PV with respective voltage is zero, negative on right of MPP and positive on left of MPP, on the P vs. *V* curve as expressed in Equation (5)

$$
\frac{dP}{dV} = \frac{d(IV)}{dV} \approx I + V \frac{\Delta I}{\Delta V}
$$
\n(5)

In INC method ∆*V* term play important role to decide the tracking speed [\[40\]](#page-18-2). The flow chart of In INC method Δ*V* term play important role to decide the tracking speed [40]. The flow chart of INC method is shown in Figure [6.](#page-6-0) INC method is shown in Figure 6.

Figure 6. Flowchart of incremental conductance (INC) method. **Figure 6.** Flowchart of incremental conductance (INC) method.

3.4. Fuzzy Logic Based MPPT 3.4. Fuzzy Logic Based MPPT

The advantage of fuzzy logic is without dealing with nonlinearity and without precise The advantage of fuzzy logic is without dealing with nonlinearity and without precise mathematical model, it can work with inaccurate input. Fuzzification, rule base table and mathematical model, it can work with inaccurate input. Fuzzification, rule base table and defuzzification, these are the steps involve in fuzzy logic. The input for fuzzy logic are error signal defuzzification, these are the steps involve in fuzzy logic. The input for fuzzy logic are error signal and change in *CE* which is given by Equations (6) and (7) and output is duty cycle. and change in *CE* which is given by Equations (6) and (7) and output is duty cycle.

$$
E(k) = \frac{P_{ph}(k) - P_{ph}(k-1)}{i_{ph}(k) - i_{ph}(k-1)}
$$
\n(6)

$$
CE(k) = E(k) - E(k-1)
$$
\n(7)

The accuracy of fuzzy logic depends on the number of fuzzy level involved in algorithm. In The accuracy of fuzzy logic depends on the number of fuzzy level involved in algorithm. In algorithm, first $E(k)$ and $CE(k)$ is calculated. Based on these values, a rule based table is formed. The last stage in fuzzy algorithm is defuzzification in which linguistic variable is converter to numerical value with the help of membership function.

4. Particle Swarm Optimization (PSO) Algorithm 4. Particle Swarm Optimization (PSO) Algorithm

The PSO is an evolutionary algorithm that gives a better result with every iteration. It is based The PSO is an evolutionary algorithm that gives a better result with every iteration. It is also
not flock concept where each bird follows the leader. In this case, each particle follows the best possible particle. In conventional PSO, there is a population of particles. The positions of these particles particles presence parameters in conventional PSO, there is a population of particles. The position of these particles are compared with the local best position and the global best position, and accordingly these particles are compared with the compared with the position and the ground to position, and accordingly after particle in
are moved in search space. The position of a particle is determined by the best particle in a region P_{best-i} as well as the best solution found by all the particles in the total population G_{best-i} . Movement of particles in the optimization process is shown in Figure [7](#page-7-1) [\[8,](#page-16-4)[20\]](#page-17-3). in a region *Pbest-i* as well as the best solution found by all the particles in the particles in the population found by a particles in the particles in

The particle position x_i is adjusted using Equation (8) The particle position μ is adjusted using α indicent (8)

$$
x_i^{k+1} = x_i^k + \phi_i^{k+1}
$$
 (8)

where the velocity component ϕ_i represents the step size. The velocity is calculated by Equation (9)

$$
\phi_i^{k+1} = \omega \phi_i^k + c_1 r_1 \left\{ P_{besti} - x_i^k \right\} + c_2 r_2 \left\{ G_{besti} - x_i^k \right\} \tag{9}
$$

where ω is the inertia weight, the acceleration coefficients are represented as c_1 and c_2 , the random constants r_1 , $r_2 \in U(0, 1)$, P_{besti} and G_{besti} is the personal best position of particle *i*, and is the best position of the particles in the entire population respectively. Since the PSO algorithm is used to find position of the particles in the entire population respectively. Since the PSO algorithm is used to find the best duty cycle for switching the MOSFET of the ZSI. The position can be taken as duty cycle, the best duty cycle for switching the MOSFET of the ZSI. The position can be taken as duty cycle, and then Equation (10) can be rewritten as and then Equation (10) can be rewritten as

$$
d_i^{k+1} = d_i^k + \phi_i^{k+1}
$$
 (10)

Figure 7. Movement of particles in the optimization process. **Figure 7.** Movement of particles in the optimization process.

In the PSO Algorithm, first the population of particles is formed and then the objective function In the PSO Algorithm, first the population of particles is formed and then the objective function is formed. Since it is implemented for Maximum Power Point Tracking (MPPT), the objective is to is formed. Since it is implemented for Maximum Power Point Tracking (MPPT), the objective is to maximize the output power. The objective function of PSO is framed as $max((P(k), P(k-1))$. In this work, the PSO Algorithm is implemented as a closed loop system for reducing the steady $\frac{1}{2}$. oscillations and improving the tracking time to achieve maximum power. While forming the state oscillations and improving the tracking time to achieve maximum power. While forming the population of particles, it is necessary to bound them to a specific level. If not, the particles will lose population of particles, it is necessary to bound them to a specific level. If not, the particles will lose their directions leading to the failure of the algorithm. The variation of the duty cycle is shown in their directions leading to the failure of the algorithm. The variation of the duty cycle is shown in Figure 8 with switching frequency of 10 kHz. Figure 9 represents the PSO based MPPT integrated Figure [8](#page-8-0) with switching frequency of 10 kHz. Figure [9](#page-8-1) represents the PSO based MPPT integrated through Z-Source inverter and Figure 10 represents the flowchart of PSO algorithm. According to through Z-Source inverter and Figure [10](#page-9-1) represents the flowchart of PSO algorithm. According to Equation (14), the duty cycle is restricted to 0 to 0.5. The constraint for duty cycle is 0.1 to 0.4 is Equation (14), the duty cycle is restricted to 0 to 0.5. The constraint for duty cycle is 0.1 to 0.4 is considered in the PSO algorithm. considered in the PSO algorithm.

Figure 8. Varying duty cycle of Particle Swarm Optimization (PSO) algorithm. initialization, the change in duty cycle is large, then the particles will have large search area. So,

When slow variation in irradiance, changes in duty cycle are small to track the MPP. Due to initialization, the change in duty cycle is large, then the particles will have large search area. So, some energy is wasted during the exploration process. If the above changes occur frequently, then the tracking needs to be fast to follow the change, but does not follow the MPP accurately. To overcome this issue modified PSO is used which is detailed in [\[20\]](#page-17-3).

In modified PSO algorithm, the G_{best} value is adjusted instead of varying the PSO parameter ω , C_1 , C_2 . The value of $C_1 = C_2 = 2F$, $\omega = 0.4$. In slowly varying irradiance, the duty cycles are initialized in two phases. First, the duty cycle is adjusted by k_1 , using Equation (11) then perturbation of two extreme two phases. First, the duty cycle is adjusted by k_1 , using Equation (11) then perturbation of two
duty cycle with a constant value of k_2 , in both the direction shown in Equation (13). (11)

$$
d_{new} = d_{old} - \frac{1}{k_1} (P_{oldMPP} - P_{MPP})
$$
\n(11)

$$
k_1 = \frac{\Delta P_{MPP}}{\Delta d}, k_1 = \{k_1 \text{ if } \Delta P > 0
$$

=
$$
\left\{\frac{k_1}{2} \text{ if } \Delta P < 0\right\}
$$
 (12)

$$
d_{i,new}^k = [d_1 - k_2, d_2, d_3 + k_2] \text{ for } k_2 \ge 0.05
$$
 (13)

Figure 9. Closed loop model of proposed PSO method.

Figure 10. Flowchart of PSO algorithm. **Figure 10.** Flowchart of PSO algorithm.

5. Z-Source Inverter 5. Z-Source Inverter

The split-inductor and capacitors are connected in 'X' shape to form the two-port impedance The split-inductor and capacitors are connected in 'X' shape to form the two-port impedance source. Impedance source inverter, which uses a unique impedance network connected between source. Impedance source inverter, which uses a unique impedance network connected between power source and an inverter circuit. It provides both voltage buck and boosts operations when compared with the traditional inverters. DC source can be a battery, fuel cell, PV cell, or output from the rectifier. the rectifier.

The circuit shown in Figure [11](#page-10-0) has six switches with antiparallel diodes are used to provide
The circuit shown in Figure 11 has six switches with antiparallel diodes are used to provide bidirectional current flow and unidirectional voltage blocking capability [19–22]. bidirectional current flow and unidirectional voltage blocking capability [\[19–](#page-17-2)[22\]](#page-17-10).

Figure 11. Circuit diagram of ZSI.

5.1. Non-Shoot through State 5.1. Non-Shoot through State **D**

In non-shoot through mode, the inverter is functioning in any one of the six traditional active In non-shoot through mode, the inverter is functioning in any one of the six traditional active vectors; the equivalent circuit is as exposed in Figure [12a](#page-10-1). vectors; the equivalent circuit is as exposed in Figure 12a. **T1 L1 T3 T5**

$$
V_{c1} + V_{c2} > V_{pv}
$$
 (14)

$$
V_{L1} = V_{c1} V_{L2} = V_{c2}
$$
 (15)

The inverter acts as a current source when observed from the DC link. As shown in Figure 12a.
the current in the inductor L_1 and L_2 are equal due to its symmetrical configuration. This feature as a current source when The inverter acts as a current source when observed from the DC link. As shown in Figure [12a](#page-10-1), increases the conducting interval of the inverter switches.

L2 During non-shoot through state,

$$
V_d = V_{pv} \tag{16}
$$

$$
B = \frac{1}{(1 - 2D)}\tag{17}
$$

$$
V_C = \frac{(1 - D)}{(1 - 2D)} * V_{pv}
$$
\n(18)

Figure 12. Equivalent Circuit of the ZSI in (a) Non-Shoot through States; and, (b) Shoot-Through State.

5.2. Shoot through State 5.2. Shoot through State

In the shoot through mode, the inverter operates in one of the seven shoot-through states. Here, the inductors L_1 and L_2 supply the inverter. The energy stored in the inductors during non-shoot Here, the inductors during non-shoot In the shoot through mode, the inverter operates in one of the seven shoot-through states. = Here, the inductors L_1 and L_2 supply the inverter. The energy stored in the inductors during non-shoot

through state is transferred to the inverter in this mode. The equivalent circuit of the inverter bridge in this mode is adjusted to the shown in Figure 12b. The time period of the shown in Figure 12b. The shown in Figure this mode is as shown in Figure [12b](#page-10-1). The time period of the shoot through state is adjusted to obtain the required value of the shoot interval is only a small interval interval in the shoot in the shoot in the shoot in the the required value of boost voltage. Generally, the shoot through interval is only a small portion of the
المقامات switching interval. $\sum_{i=1}^{n}$ shown the shock through $\sum_{i=1}^{n}$

During shoot through mode,

$$
V_d = 2V_c, V_i = 0 \tag{19}
$$

5.3. Traditional Zero State 5.3. Traditional Zero State

Figure 13 represents the equivalent circuit of the inverter in traditional zero states. Figur[e 13](#page-11-1) represents the equivalent circuit of the inverter in traditional zero states.

Figure 13. Equivalent Circuit of the ZSI in traditional zero states. **Figure 13.** Equivalent Circuit of the ZSI in traditional zero states.

The upper or lower three switches of the inverter are operating causes the traditional zero state. The upper or lower three switches of the inverter are operating causes the traditional zero state. During this time, the inverter acts as an open circuit viewed from the Z-source and inductor carries During this time, the inverter acts as an open circuit viewed from the Z-source and inductor carries current, which strengthens the line current's harmonic reduction. current, which strengthens the line current's harmonic reduction.

Switching frequency of the inverter (*fsw*) is 10 kHz. By using Equation (20), L and C values are Switching frequency of the inverter (*fsw*) is 10 kHz. By using Equation (20), L and C values are calculated and are used in hardware. Where D is shoot through duty ratio, V_C-capacitor voltage, is the average inductor current and ∆ܫ is inductor current ripple. *Iavg* is the average inductor current and ∆*I* is inductor current ripple.

$$
L = D * V_C * \frac{D * V_C}{f_{sw} * \Delta I}
$$

\n
$$
C = \frac{I_{avg} * D}{0.03 f_{sw} * V_C}
$$
\n(20)

6. Results and Discussion

Under very challenging conditions, namely partial shading of the PV array, step changes in load, and step changes in irradiance, MATLAB simulations are carried out to evaluate the effectiveness of PSO method. Figure 14a,b represent the output power from PSO based MPPT. Figure 15a,b shows the comparison of P[SO](#page-12-1) with conventional P&O algorithm. From Figures 14 and 15a the output power settled after 0.4 s in PSO algorithm, but in P&O 0.75 s. The outputs of PSO and conventional P&O methods are compared to see the effectiveness of presented PSO based MPPT. From Figure 15, PSO based MPPT has less tracking time and steady state oscillations.

Figure 14. Output power from PSO for (a) $1000W/m^2$; (b) irradiance is varied from $1000 W/m^2$ to 500 W/m² at 1 s then increased to 1000 W/m² at 2 s.

Figure 15. Comparison of output power from PSO based MPPT and P&O based MPPT for $(2.1000 \text{ W/m}^2, (h)$ invadings is varied from 1000 W $/m^2$ to 500 W (m^2) (a) 1000 W/m^2 ; (b) irradiance is varied from 1000 W/m^2 to 500 W/m^2 .

From Figures 14 and 15, PSO based MPPT produces lesser oscillations and tracking time when compared to P&O algorithm. The tracking time of PSO algorithm is 0.2 s, but for the P&O algorithm, is 0.35 s. PSO algorithm decides the shoot through duty cycle for the Z-Source inverter, according to the PV voltage. Figure 16 shows the output voltage V_L , current i_L , power P_L and shoot through duty cycle D of the Z-Source inverter. Figure [17](#page-13-1) represents the phase to phase voltage of the Z-Source inverter.

Figure 16. Output of Z-Source inverter with PSO based Maximum Power Point Tracking (MPPT).

Figure 17. Phase to phase voltage of Z-Source inverter across resistive load.

Hardware Setup Hardware Setup Hardware Setup

Figure 18 shows the hardware setup of PSO based MPPT with Z-Source inverter. Here, dspace 11gue 10 shows the hardware setup of 1 50 based will 1 while 2-50 dice inverter. Here, uspace Trox controller is used to obtain optimum auty cycle based upon the FSO algorithm and generate sensors and current are led to most a through a driver chedit. Voltage sensors and current sensors are used to sense the voltage and Current Coming none I v system. Dased upon these values best duty ζ) Let us calculated and ST vvivi pulses are generated. The WOSPET INF040D is used in ζ -Source inverter. The value of capac. Figure [18](#page-14-0) shows the hardware setup of PSO based MPPT with Z-Source inverter. Here, dspace 1104 controller is used to obtain optimum duty cycle based upon the PSO algorithm and generate 1104 controller is used to obtain optimum duty cycle based upon the PSO algorithm and generate SPWM pulses, which are fed to MOSFET through a driver circuit. Voltage sensors and current SPWM pulses, which are fed to MOSFET through a driver circuit. Voltage sensors and current sensors are used to sense the voltage and current coming from PV system. Based upon these values best duty cycle is calculated and SPWM pulses are generated. The MOSFET IRF840B is used in Z-Source inverter. The value of capacitors and inductors in the Z-Source network is 1200 μ f, 220 V, and 50 μ H, 5–6 A.

The dspace controller has two modes of operation, such as master Γ PC and slave mode for F wive generation. In slave mode, has only four channels and is used for fixed duty cycle. Implement the $\,$ tequired algorithm in master PrC shown in Figure 19. The outputs are taken from CP17 connector $\lim_{\alpha \to \infty}$ the pins 2, 3, 3, 21, 23, and 24 with respect to pin 1 (GND). The PWM generated from uspace α controller is 5 α , this 5 α is not enough to drive the power Electronic devices. So, it is boosted through t driver IC MIC4422. The PWM output from the driver is represented in Figure 20. The dspace controller has two modes of operation, such as master PPC and slave mode for PWM generation. In slave mode, has only four channels and is used for fixed duty cycle. Implement the $\frac{1}{10}$ required algorithm in master PPC shown in Figure [19.](#page-14-1) The outputs are taken from CP17 connector from the pins 2, 3, 5, 21, 23, and 24 with respect to pin 1 (GND). The PWM generated from dspace controller is 5 V, this 5 V is not enough to drive the power Electronic devices. So, it is boosted through driver IC MIC4422. The PWM output from the driver is represented in Figure [20.](#page-15-0) $\frac{1}{1}$ and $\frac{1}{2}$ becomes in Figure 20.

Figure 18. Hardware setups of PSO based MPPT with ZSI. **Figure 18.** Hardware setups of PSO based MPPT with ZSI. **Figure 18.** Hardware setups of PSO based MPPT with ZSI.

Figure 19. SPWM generation in dspace. **Figure 19.** SPWM generation in dspace.

Figure 20. SPWM generated after driver circuit.

Figure [21](#page-15-1) represents the voltage across cross coupled capacitor is 50 V, which is twice of the input supply. Output voltage across the load is shown in [Figu](#page-15-2)res [22](#page-15-3) and 23. The output peak to peak voltage is 98 V approximately equal to the simulation output shown i[n Fi](#page-13-1)gure 17. is 98 V approximately equal to the simulation output shown in Figure 17.

Figure 21. Voltage across cross-coupled capacitors of ZSI (50 V/Div).

Figure 22. Phase to phase voltage across resistive load. **Figure 22.** Phase to phase voltage across resistive load. **Figure 22.** Phase to phase voltage across resistive load.

Figure 23. Phase to phase voltage across resistive load (20 Volts/Div).

7. Conclusions

This paper involves a dspace implementation of PSO based MPPT for obtaining optimum duty cycle for the Z-Source inverter. From the obtained duty cycle, the modulation index is estimated and given to the PWM generation circuit. PWM output from the dspace controller is 5 V, which is not enough to drive the MOSFET switches, so this 5 V is increased to 12 V using a driver circuit. The PSO algorithm locates the real MPP under a change in environmental conditions, provides better tracking speed, and fewer oscillations, to improve the stability of the system. So, the PSO algorithm overcomes the shortage of conventional MPPT algorithm. Practical experiments are conducted for 60 W PV panel with different irradiance condition. This tracking algorithm is further consistent method to locating the global MPP than a conventional MPPT algorithm, predominantly extreme change in environmental conditions and shaded PV modules.

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